

## CORPORATE FED MICROSTRIP ANTENNA ARRAY WITH ENHANCED BANDWIDTH

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### ABSTRACT

The paper throws light on the potentiality of electromagnetic band gap structure and defective ground structure in ameliorating the performance of four element microstrip antenna array. Plus shape patch type electromagnetic band gap structure and rotated dumb bell defective ground structure are employed. The proposed antenna array has bandwidth and gain equal to 54.54 % and 12.77 dB respectively. Virtual size reduction of 26.58 % is also produced. The interference level is improved with reduction in mutual coupling value from -17.83dB to -25.13 dB respectively. Moreover, the proposed microstrip antenna array is also yielding appreciable reduction in back lobe radiation. The dielectric substrate employed is FR-4 glass epoxy. The microstrip antenna arrays are designed using Mentor Graphics IE3D software and measured results are obtained using vector network analyzer.

**Keywords:** Antenna Array, Bandwidth, Defective Ground Structure, Dielectric Substrate, Electromagnetic Band Gap Structure, Mutual Coupling, Radiation Pattern

### 1. INTRODUCTION

An antenna is defined as a device for radiating or receiving electromagnetic waves. It is usually required to concentrate the radiated energy in some directions and suppress it in other directions. [1-2]. In recent years wideband systems and technologies are inspiring many scientists, researchers and academicians. The antenna is a very important component of the wideband systems. In certain applications high gains are desired, which are possible using antenna arrays. During the past few years there is a huge demand for high gain, wideband and compact antennas. Microstrip antennas are low profile antennas with simple structure. A patch antenna consists of a metal piece on top of a dielectric substrate. The whole assembly is mounted on a finite ground plane. One of the main limitations of microstrip antenna arrays is the high values of mutual coupling between the antenna elements. Researchers are facing lot of hardships in controlling the drawback of microstrip antenna arrays. [3-4].

Electromagnetic band gap (EBG) structures are artificial periodic structures which aid in improving the mutual coupling levels of microstrip antenna arrays. They are defined as structures which stop or assist the propagation of electromagnetic waves in certain frequency bands. Defective ground structures (DGS) are structures which are loaded in the ground plane of microstrip antenna arrays. [5].

In [6] Eli Yablonovitch has exhibited that if a three dimensional periodic dielectric structure has an electromagnetic band gap which overlaps the electronic band edge, then spontaneous emission can be rigorously suppressed and also stimulated emission would be absent. There is periodic modulation along the laser axis and therefore a forbidden gap in the electromagnetic dispersion relation. In [7] D. N. Elsheakh et al have discussed the study of EBG structures loaded in the ground plane, their types, and their behaviour in enhancing the performance of two element microstrip patch antenna arrays. The EBG structures employed are of two dimensional in nature and

corporate feeding technique is used to feed the antenna array. The performances of square, circular, star, H and I shape EBG structures are compared. Highest bandwidth of 5.1 % has been achieved using H shape EBG structure. Least amount of mutual coupling (S21) of -30 dB and highest gain of 13.75 dB have been obtained in the case of Star EBG structure. In [8] Vani H. R et al have designed a novel microstrip antenna array with split ring resonator metamaterial loaded in the ground plane. A single layer of split ring resonator is placed around the patch. The 1 mm gap in the ring gives capacitive effect to split ring resonator. The resonant frequency of patch 1 surrounded by split ring resonator is shifted from 6.8 to 5.5 GHz with a bandwidth of 21.2 %. The resonant frequency of patch 2 is shifted from 6.8 to 6.4 GHz with a bandwidth of 6.25 %. The coupling between the patches separated by 7 mm is decreased by 10 dB in the frequency range 6 to 7.5 GHz. When a slot of dimensions of 3.5 mm × 0.3 mm is placed on patch 1, the resonant frequency is shifted to 6 GHz. Hence miniaturizations of 42 and 27.5 % of patch 1 and patch 2 are produced. In [9] Hassan Elesway et al have designed and fabricated single, two and four element linear microstrip antenna arrays (E shape) using corporate feeding technique. Two parallel slots are cut to perturb the surface current path and enhance the bandwidth. The inter element spacing is 13.6 mm. The directivities of single, two and four element antenna arrays are 7, 9 and 12 dBi respectively. The corresponding maximum achievable gains are equal to 5, 7.5 and 11 dBi respectively. Bandwidths obtained are equal to 8, 11 and 16 % respectively. The maximum values of antenna and radiation efficiencies are 80 and 88 %. In [10] Ali Hanafiah Rambe et al have improved the performance of microstrip antenna using a stub connected to the microstrip feed line. After the introduction of the stub, the return loss is improved by 139 %. The values of bandwidths without and with stub are equal to 62.1 and 70.8 MHz respectively for  $VSWR \leq 2$ . Additionally, there is 2 % increment in the value of gain. In [11] H. Errifi et al have compared the performance of microstrip antenna array using series, corporate and series – corporate feed methods.

The dielectric substrate employed is RT – Duroid. Microstrip antenna arrays consisting of two, four and eight elements have been designed. The distance between the adjacent antenna elements is  $\lambda/2$ . In the case of two element antenna array, corporate feeding method is the best method as it is producing highest gain and directivity of 9.93 and 10.19 dB respectively. In the case of four element antenna array, corporate feeding method is the best method as it is producing highest gain and directivity of 12.85 and 13.18 dB respectively. In the case of eight element antenna array, series - corporate feeding method is the best method as it is producing highest gain and directivity of 14.13 and 14.51 dB respectively. In [12] Shruti Dhamankar et al have discussed the different techniques employed for mutual coupling reduction in microstrip antenna arrays. These are using EBG structures, split ring resonators, EBG – split ring resonators and DGS. Using EBG and split ring resonator between two radiating elements, reduces the mutual coupling more compared to other techniques. The mutual coupling is reduced to more than 42 dB and side lobes by 6 dB. The proposed antenna with spiral resonator has 5.5 dB reduction in mutual coupling. A 36 dB reduction in mutual coupling is achieved using DGS. In [13] F. Benikhlef et al have exhibited the reduction of mutual coupling using mushroom like EBG structure on the surface of two element microstrip antenna array. Initially the size of the EBG cells is varied. With the dimensions of unit cells as  $3.5\text{ mm} \times 3.5\text{ mm}$  and periodicity as 1 mm, mutual coupling is reduced from -32.07 to -45.72 dB. For the same dimensions of unit cells the periodicity of unit cells is varied. For periodicity of 0.8 mm, best reduction in mutual coupling to -41.39 dB is obtained. The same process is repeated with square Koch fractal as radiating patch. The separation between the patches is 22.5 mm. Highest reduction in mutual coupling is observed with one row of EBG equal to -51.2 dB. Good amount of gain of 7.4 dB is also obtained. In [14] Sandhya Bhavsar et al have discussed the various types and their behaviour in enhancing the performance characteristics of microstrip antennas. It is found that EBG structures can alleviate the effect of surface wave radiation and thereby decreasing the mutual coupling between the antenna elements. The effect of EBG structures on other parameters like antenna gain, radiation pattern, bandwidth and beam width is also demonstrated. In [15] Mohammad Naser - Moghadasi et al have designed  $2 \times 5$  EBG structure to reduce mutual coupling between patch antennas of MIMO array. Two microstrip patch antennas are designed for resonance at 5.28 GHz. The conventional MIMO array is fed by coaxial feed and bandwidth is equal to 3 %. The EBG structure is inserted between the two patch antennas and on the surface. Mutual coupling values without and with EBG structure are -22 and -43 dB respectively. By increasing the gap between the unit cells of EBG structure, the resonant frequency of proposed MIMO array is reduced. A gain value of 6.86 dBi is also produced. Moreover the EBG structure has reduced antenna current from  $8.5\text{ A/m}$  to  $3.9\text{ A/m}$ , so the coupling is reduced by 50 %. However the antenna efficiency is reduced from 65 to 53 %. In [16] Duong Thi Thanh Tu et al have designed dual band MIMO antenna system with enhanced isolation. Using a double rectangular DGS, the antenna is resonating at 2.6 and 5.7 GHz with bandwidths of 5.7 and 4.3 % respectively. The proposed antenna is

having high isolation which is stable and around -20 dB over all frequencies. At 2.6 GHz, gain and radiation efficiency are 2.63 dB and 59 %. The corresponding values at 5.7 GHz are 1.6 dB and 39.8 %. MIMO antenna with double side EBG structure is reducing mutual coupling from -20 to -40 dB. At 2.6 GHz the antenna gain and radiation efficiency are improved to 4.25 dB and 68.7 %. At 5.7 GHz, the antenna gain is increased to 1.76 dB and radiation efficiency to 39.8 %. In [17] Zulkifli et al have employed dumbbell shaped DGS on the opposite side of the feed line. The dielectric material used has a dielectric constant of 2.2 and loss tangent of 0.0009. The antenna has two substrate layers with electromagnetic coupling feeding technique. The upper substrate layer consists of two element patch antenna with two U shape and S shape patches. The lower substrate layer consists of four dumbbell slots inserted in the ground plane. Multiband characteristics with three resonant frequencies are obtained. Bandwidths without and with DGS are almost same. However with the improvement of return loss provides an indication of increase of antenna efficiency. At 2.3 GHz, a decrease in mutual coupling of 2 to 5 dB, at 3.3 GHz, a decrease in mutual coupling of 0.6 to 3.3 dB and at 5.8 GHz, a decrease in mutual coupling to 2.9 dB are obtained. For 2.3, 3.3 and 5.8 GHz bands the gain enhancements to 2.4, 2.4 and 3.3 dB are achieved. In [18] M. Tecpoyotl - Torres et al have implemented three types of DGS for single element and antenna array of 5 and 9 patches. Using first DGS, the single element has return loss and gain equal to -21.44 dB and -0.742 dB at the resonant frequency of 2.42 GHz. For the antenna array based on five patches, return loss and gain are equal to -17.6 dB and 4.9 dB respectively. Using nine patches the corresponding values are -12 dB and 5.19 dB. The nine patch antenna array with second DGS produces enhanced gain of 7.01 dB. Finally with third DGS, a healthy gain of 15.33 dB is obtained along with a return loss of -23.24 dB. In [19] K. S. Ahmad et al have proposed the design of planar array antenna using I shaped DGS inserted between the patches. It is observed that the resonant frequency of the antenna reduces from 11.22 to 8.25 GHz, as the dimensions of DGS are varied. Significant effect on steering of the main beam is observed by changing the dimensions of the DGS. The maximum directivity produced is 6.32 dBi. The influence of DGS slot is also visible in reducing the side lobes. In [20] Otman Ouhaj et al have presented an improved method of size reduction with circular polarization using dumbbell shaped DGS. The DGS is etched in the ground plane below the microstrip line. When the DGS is introduced, the resonant frequency of the antenna array is shifted from 10 to 7.5 GHz, thus producing a miniaturization of 43.75 %. At 7.5 GHz the gain is equal to 8.99 dB. In the proposed antenna array, the current density is more concentrated along the DGS in the ground plane and structure array elements on the top face. In [21] R. A. Pandhare et al have employed U shaped dumbbell DGS to shift the resonant frequency of the microstrip antenna array from 5.2 to 3.8 GHz to obtain size reduction of 45 %. A healthy return loss of -29 dB is obtained at the resonant frequency of 3.8 GHz. However the gain of the antenna is reduced to 5.9 dB. The decrease in gain is due to the increase of lateral and longitudinal radiations due to the propagation of surface waves. The designed antenna array is useful for WiMax

application. In [22] Ali Farahbakhsh et al have designed a microstrip antenna array in which the slot is of polygon type and shape is obtained by using an enhanced genetic algorithm. The patches are fed by proximity coupled feeding method. The reflection and coupling of the proposed antenna are 3 and 4 dB respectively. The gain of the antenna with DGS is improved. The dip in the gain graph shows that the radiation pattern of the antenna is dependent on frequency and the directivity of the antenna is decreased. The E-plane coupling of the array antenna is 11 dB better than that without DGS. However the back lobe level is increased in array antenna with DGS slot and is due to ground slots. In [23] Susanta Kumar Parui et al have investigated the effect of two U shaped slots connected by a thin transverse slot placed in the ground plane of microstrip antenna. The simulated S parameters of DGS depict one pole low pass filter characteristics with an attenuation pole at 8.5 GHz and 3 dB cut off frequency at 3 GHz. Two units of DGS when considered provide a sharpness factor of 15 dB/GHz. A maximum attenuation of 33 dB is also obtained. Also as the width of the transverse slot is decreased, the stop band centre frequency also decreases due to the increment of the lumped capacitor. Additionally, as the separation between the DGS units increases, the 3 dB cut off frequency and stop band centre frequency decrease due to the increase of lumped inductance. The values of mutual coupling are also decreased.

The newness in this paper is that EBG structures are employed on the surface instead in the ground plane. Additionally, the combined effect of EBG and DGS structures on microstrip antenna arrays is observed. Comparing the published research results with the research results in this paper, highest bandwidth of 54.54 % is obtained in this paper. Moreover better reduction in back lobe radiation is also produced in this paper compared to other published results.

## 2. CONVENTIONAL MICROSTRIP ANTENNA ARRAY

The two element conventional microstrip antenna array (TCMAA) consists of two identical rectangular radiating elements designed at 6 GHz. The dielectric substrate employed is FR-4 glass epoxy which has a dielectric constant of 4.2 and loss tangent of 0.0245. The substrate height is equal to 1.6 mm. The dimensions of the finite ground plane to design TCMAA are 115.8 mm × 62.7 mm. The two elements of TCMAA are separated by a distance of  $\lambda/4$ , where  $\lambda$  is the wavelength calculated at the design frequency of 6 GHz. The feeding method employed to excite TCMAA is corporate feeding method. The dimensions of each of the radiating patches are 15.73 mm × 11.76 mm. The schematic of TCMAA is depicted in Fig. 1 and is employed to determine the return loss of CMAA. Return loss is designated by the S-parameter  $S_{11}$ .

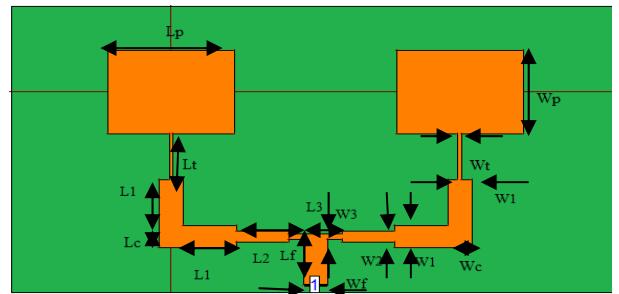


Fig. 1 CMAA

All the dimension values of the parts of CMAA are presented in Table 1.

Table 1 Dimensions and values of CMAA

| Dimension                                   | Value (mm) |
|---|------------|
| Length of the patch (Lp)                    | 15.73      |
| Width of the patch (Wp)                     | 11.76      |
| Length of the quarter wave transformer (Lt) | 6.47       |
| Width of the quarter wave transformer (Wt)  | 0.47       |
| Length of the 50Ω line (L1)                 | 6.52       |
| Width of the 50Ω line (W1)                  | 3.05       |
| Length of the coupler (Lc)                  | 3.05       |
| Width of the coupler (Wc)                   | 3.05       |
| Length of the 70Ω line (L2)                 | 6.54       |
| Width of the 70Ω line (W2)                  | 1.62       |
| Length of the 100Ω line (L3)                | 6.56       |
| Width of the 100Ω line (W3)                 | 0.70       |
| Length of the feed line (Lf)                | 6.52       |
| Width of the feed line (Wf)                 | 3.05       |

The two radiating patches of TCMAA are fed separately as shown in Fig. 2 to measure the mutual coupling between the two radiating patches. The two radiating patches are separated by the same distance as that maintained in Fig. 1. Moreover the two radiating patches are assumed to be excited with the same amount of power. Mutual coupling is designated by the S-parameter  $S_{21}$ .

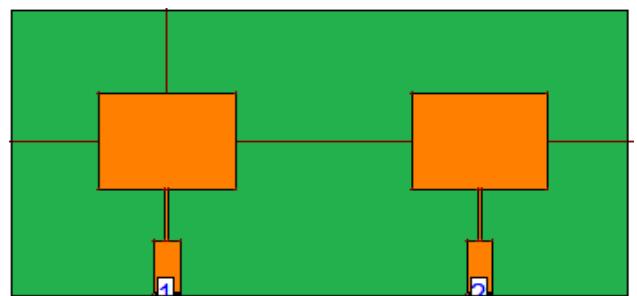
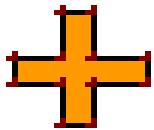


Fig. 2 Setup of CMAA for mutual coupling measurement

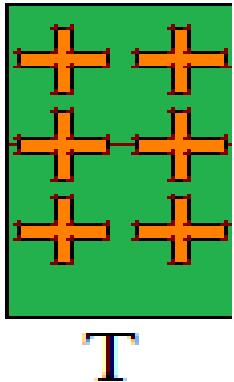
### 3. EBG AND DGS STRUCTURES

The EBG structure employed in the design consists of a matrix of 3 rows and 2 columns of plus shape patch type EBG structure. The unit cell of the EBG structure employed is depicted in Fig. 3.



**Fig. 3** Unit cell of EBG structure

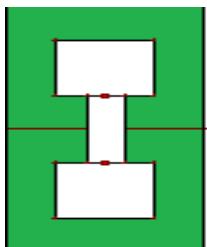
The unit cell of the EBG structure depicted in Fig. 4 consists of two rectangles intersecting each other at right angles and at their centres. The dimensions of each of the rectangles are 5 mm and 1 mm respectively. The plus shape patch type EBG structure is shown in Fig. 4.



**Fig. 4** EBG structure

In Fig. 4, the unit cells of the EBG structure are separated by a distance of  $T = 1.5$  mm along the x-axis and y-axis.

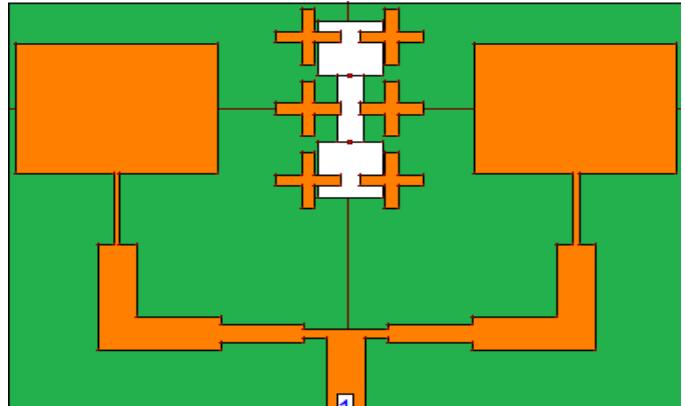
The DGS consists of a rotated dumbbell shape structure. The dumbbell structure is rotated by  $90^\circ$  from its usual horizontal position. Fig. 5 shows the schematic of DGS structure. It consists of two squares which are joined at the centres of one of their sides by a rectangle. The dimensions of each of the squares of the DGS structure are  $5 \text{ mm} \times 5 \text{ mm}$  respectively. The length and breadth of each of the rectangles of DGS structure are 6 mm and 2 mm respectively.



**Fig. 5** Schematic of DGS

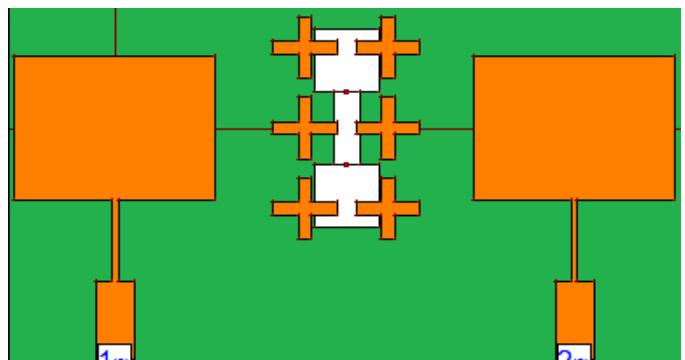
### 4. PROPOSED MICROSTRIP ANTENNA ARRAY

The design of two element proposed microstrip antenna array (TPMAA) is obtained by modifying the TCMAA. TPMAA has plus shape patch type EBG structure on its surface and rotated dumbbell DGS in its ground plane. The ground plane dimensions of TCMAA and TPMAA are same. In addition, the plus shape patch type EBG structure and rotated dumbbell DGS are loaded in between the radiating patches. The schematic of TPMAA is depicted Fig. 6. The schematic in Fig. 6 is used to measure the return loss characteristics of TPMAA.



**Fig. 6** PMAA

Fig. 7 depicts the schematic employed to determine the effect of EBG and DGS structures on the mutual coupling coefficients. It consists of setup of elements of TPMAA which are fed separately.



**Fig. 7** Setup of PMAA for mutual coupling measurement

Figs. 8, 9, 10 and 11 depict the photographs of fabricated antennas TCMAA and TPMAA.

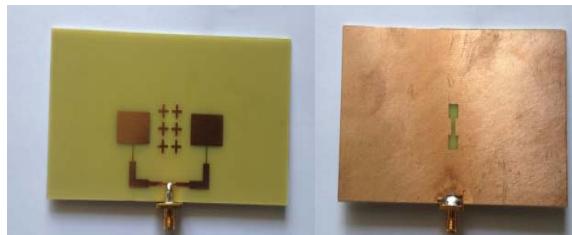


**Fig. 8** Photograph of CMAA  
(a) Front view (b) Back view

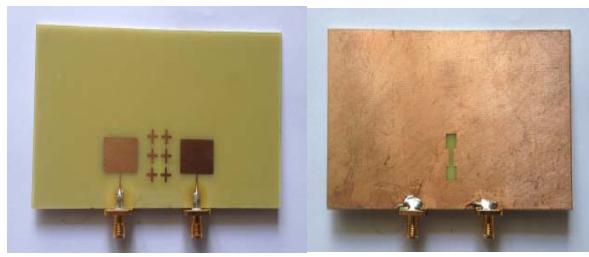


**Fig. 9** Photograph of setup of CMAA for mutual coupling measurement

(a) Front view (b) Back view



**Fig. 10** Photograph of PMAA  
(a) Front view (b) Back view

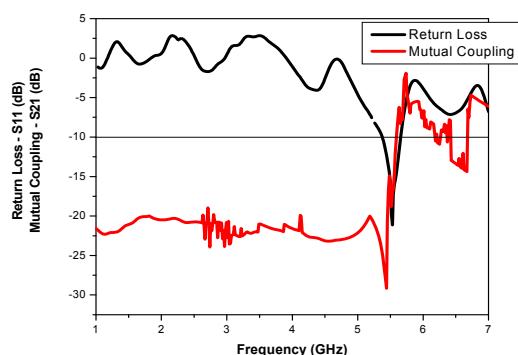


**Fig. 11** Photograph of setup of PMAA for mutual coupling measurement

(a) Front view (b) Back view

## 5. RESULTS

The measured results of the fabricated antennas TCMAA and TPMAA are obtained using vector network analyzer. Fig. 12 depicts the graph of measured results of return loss and mutual coupling versus frequency of TCMAA. The frequency range over which the measured results are taken is 1 – 7 GHz.



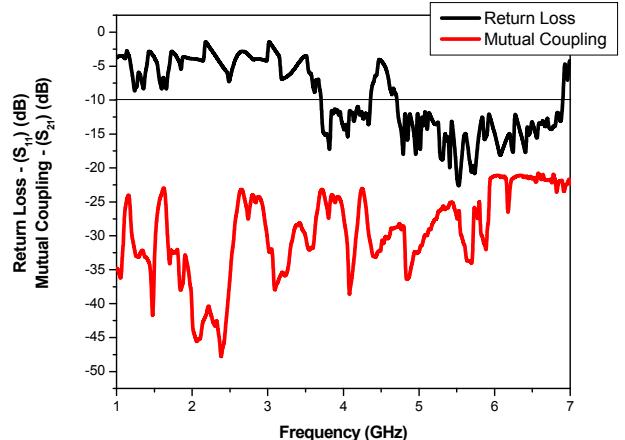
**Fig. 12** Graph of return loss and mutual coupling versus frequency of CMAA

Fig. 12 depicts that TCMAA is resonating at the fundamental frequency of 5.53 GHz with a return loss of -21.23 dB. TCMAA is producing a bandwidth of 130 GHz. Bandwidth (%) is determined by using equation (1)

$$\frac{\text{Bandwidth}}{\text{Resonant frequency}} \times 100\% \quad (1)$$

Hence bandwidth (%) of TCMAA is equal to 2.35 %. The mutual coupling parameter measured is  $S_{21}$ . From Figure 12 the value of the mutual coupling parameter at the resonant frequency of 5.53 GHz is -17.83 dB. The mutual coupling value is high as it is greater than -20 dB. Furthermore Figure 12 also shows that the plots of return loss and mutual coupling are crossing each other at the resonant frequency of 5.53 GHz. This implies there is huge amount of interference between the transmitting element 1 and the receiving element 2 of TCMAA. Therefore there is improper transmission and reception of electromagnetic signals in TCMAA.

Fig. 13 shows the graph of measured return loss and mutual coupling characteristics versus frequency of of TPMAA.



**Fig. 13** Graph of return loss and mutual coupling versus frequency of PMAA

From Fig. 13 we see that TPMAA is resonating at dual bands i.e. at 3.81 and 5.53 GHz respectively. FPMAA is producing bandwidths of 625 and 2110 MHz at these two resonant frequencies. Thus TPMAA is producing an overall bandwidth of 54.54 %, which is greater than 2.35 % produced by its counterpart i.e. TCMAA. The value of mutual coupling obtained at the resonant frequency of 5.53 GHz is equal to -25.13 dB. The value of mutual coupling parameter is reduced compared to that obtained by TCMAA. Also, the graphs of return loss and mutual coupling of TPMAA are not overlapping with each other at the resonant frequency of 5.53 GHz. This confirms the decreased interference level in TPMAA compared to that in TCMAA. Hence TPMAA is a better performer than TCMAA in terms of parameters bandwidth (%) and mutual coupling.

Figs. 12 and 13 depict that TCMAA and TPMAA are producing fundamental resonant frequencies of 5.53 and 3.81 GHz respectively. TPMAA is thus resonating at lower fundamental frequency compared to TCMAA. This accounts to virtual size reduction. To calculate the virtual size reduction (%) produced by TPMAA equation (2) is employed.

$$\left(\frac{f_1-f_2}{f_1}\right) \times 100 \quad (2)$$

In equation (2)  $f_1$  and  $f_2$  are the fundamental resonant frequencies of TCMAA and TPMAA. Substituting the relevant parameters in equation (2), the virtual size reduction (%) produced by TPMAA is calculated as equal to 26.58 %.

The gains of the antennas TCMAA and TPMAA are determined by using the equation (3)

$$G = 20 \log_{10} \left( \frac{4\pi R}{\lambda} \right) + 10 \log_{10} \frac{Pr}{Pt} - Gt \quad (3)$$

where:  $Pt$  – transmitted power,  $Pr$  – received power,  $R$  – distance between the transmitting and the receiving antennas,  $\lambda$  – wavelength at the resonant frequency of 5.53 GHz,  $Gt$  – gain of the transmitting antenna.

$Gt$  is given by the equation (4)

$$Gt = 10 \log_{10} Gs \quad (4)$$

$$Gs = \frac{2\pi ab}{\lambda^2} \quad (5)$$

where  $a$  and  $b$  are the length and width of the standard pyramidal horn antenna used as the transmitting antenna. The receiving antenna is the antenna under test i.e. TCMAA and TPMAA respectively. The dimensions  $a$  and  $b$  are equal to 24 and 14 cm respectively. The distance between the transmitting antenna (standard pyramidal horn antenna) and the receiving antenna is given by the equation (6)

$$R \geq \frac{2D^2}{\lambda} \quad (6)$$

where  $D$  is the larger dimension of the transmitting antenna. The dimension  $D$  is equal to 24cm. The value of  $R$  is equal to 71.86m.

Initially considering TCMAA as the receiving antenna, the transmitted and received powers of TCMAA are equal to 8.7  $\mu$ W and 8.8 nW respectively. Substituting the corresponding parameter values in equation (3), TCMAA is producing gain equal to 5.069 dB. Considering TPMAA as the receiving antenna, the measured values of transmitted and received powers are equal to 8.7  $\mu$ W and 0.0489  $\mu$ W respectively. Substituting the corresponding parameter values in equation (3), the value of gain of TPMAA is equal to 12.77 dB. Thus TPMAA is producing enhanced gain of 12.77 dB as compared to 5.069 dB produced by TCMAA. Hence TPMAA is a better antenna than TCMAA in terms of gain parameter.

The radiation pattern of an antenna is obtained by plotting the values of electromagnetic power at various angles from  $0^\circ$  to  $360^\circ$  on a polar plot. The forward power is measured at the angle of  $90^\circ$  and back lobe radiation at the angle of  $360^\circ$ . The radiation plots of the conventional antenna array TCMAA and the proposed antenna array TPMAA are depicted in Fig. 14.

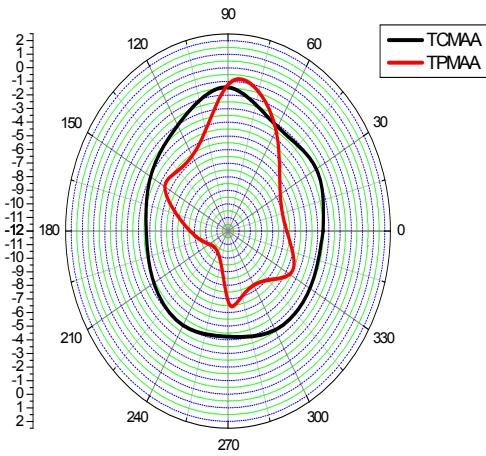


Fig. 14 Plot of radiation patterns of CMAA and PMAA

The radiation pattern of TCMAA depicts that TCMAA is radiating a forward power of -1.31 dB and a back lobe radiation of -4.18 dB respectively. In the presence of rotated dumbbell shape DGS and plus shape patch type EBG structure, TPMAA is producing increased forward power of -1 dB and reduced backward power of -7 dB respectively. The front to back ratio (FBR) parameter is calculated by subtracting the power radiated in the backward direction from the power radiated in the forward direction. Therefore, the FBR values of TCMAA and TPMAA are equal to 2.87 and 6 dB respectively. Comparing TCMAA and TPMAA in terms of forward power, backward power and FBR values, TPMAA is radiating enhanced power in the desired direction and decreased power in the undesired direction. Hence, TPMAA is a better performer than TCMAA as far as forward and backward powers are concerned. The greater value of FBR of TPMAA than that produced by TCMAA indicates that TPMAA is a better radiator than TCMAA in terms of FBR.

## 6. CONCLUSIONS

The conventional and proposed microstrip antenna arrays have been successfully designed, fabricated and tested experimentally. A healthy bandwidth of 54.54 % and appreciable reduction in mutual coupling are produced, indicating the ability of plus shape patch type electromagnetic band gap structure and rotated dumbbell defective ground structure. Additionally, the gain value is enhanced to 12.77 dB. The better performance of proposed microstrip antenna array is demonstrated by a good value of virtual size reduction and notable decrease in back lobe radiation.

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